

Problem Solutions for Chapter 6

6-1. From Eqs. (6-4) and (6-5) with $R_f = 0$, $\eta = 1 - \exp(-\alpha_s w)$

To assist in making the plots, from Fig. P6-1, we have the following representative values of the absorption coefficient:

| λ (μm) | α_s (cm^{-1}) |
|-----------------------------|---------------------------------|
| .60 | 4.4×10^3 |
| .65 | 2.9×10^3 |
| .70 | 2.0×10^3 |
| .75 | 1.4×10^3 |
| .80 | 0.97×10^3 |
| .85 | 630 |
| .90 | 370 |
| .95 | 190 |
| 1.00 | 70 |

$$\begin{aligned}
 6-2. \quad I_p &= qA \int_0^w G(x) dx = qA \Phi_0 \alpha_s \int_0^w e^{-\alpha_s x} dx \\
 &= qA \Phi_0 \left[1 - e^{-\alpha_s w} \right] = qA \frac{P_0(1 - R_f)}{h\nu A} \left[1 - e^{-\alpha_s w} \right]
 \end{aligned}$$

$$6-3. \quad \text{From Eq. (6-6),} \quad R = \frac{\eta q}{h\nu} = \frac{\eta q \lambda}{hc} = 0.8044 \eta \lambda \quad (\text{in } \mu\text{m})$$

Plot R as a function of wavelength.

6-4. (a) Using the fact that $V_a \approx V_B$, rewrite the denominator as

$$\begin{aligned}
 1 - \left(\frac{V_a - I_M R_M}{V_B} \right)^n &= 1 - \left(\frac{V_B - V_B + V_a - I_M R_M}{V_B} \right)^n \\
 &= 1 - \left(1 - \frac{V_B - V_a + I_M R_M}{V_B} \right)^n
 \end{aligned}$$

Since $\frac{V_B - V_a + I_M R_M}{V_B} \ll 1$, we can expand the term in parenthesis:

$$1 - \left(1 - \frac{V_B - V_a + I_M R_M}{V_B}\right)^n \approx 1 - \left[1 - \frac{n(V_B - V_a + I_M R_M)}{V_B}\right]$$

$$= \frac{n(V_B - V_a + I_M R_M)}{V_B} \approx \frac{n I_M R_M}{V_B}$$

Therefore, $M_0 = \frac{I_M}{I_p} \approx \frac{V_B}{n(V_B - V_a + I_M R_M)} \approx \frac{V_B}{n I_M R_M}$

(b) $M_0 = \frac{I_M}{I_p} = \frac{V_B}{n I_M R_M}$ implies $I_M^2 = \frac{I_p V_B}{n R_M}$, so that $M_0 = \left(\frac{V_B}{n I_p R_M}\right)^{1/2}$

6-5. $\langle i_s^2(t) \rangle = \frac{1}{T} \int_0^T i_s^2(t) dt = \frac{\omega}{2\pi} \int_0^{2\pi/\omega} R_0^2 P^2(t) dt$ (where $T = 2\pi/\omega$),

$$= \frac{\omega}{2\pi} R_0^2 \int_0^{2\pi/\omega} (1 + 2m \cos \omega t + m^2 \cos^2 \omega t) dt$$

Using

$$\int_0^{2\pi/\omega} \cos \omega t dt = \frac{1}{\omega} \sin \omega t \Big|_{t=0}^{t=2\pi/\omega} = 0$$

and $\int_0^{2\pi/\omega} \cos^2 \omega t dt = \frac{1}{\omega} \int_0^{2\pi} \left(\frac{1}{2} + \frac{1}{2} \cos 2x\right) dx = \frac{\pi}{\omega}$

we have $\langle i_s^2(t) \rangle = R_0^2 P_0^2 \left(1 + \frac{m^2}{2}\right)$

6-6. Same problem as Example 6-6: compare Eqs. (6-13), (6-14), and (6-17).

(a) First from Eq. (6-6), $I_p = \frac{\eta q \lambda}{hc} P_0 = 0.593 \mu A$

Then $\sigma_Q^2 = 2q I_p B = 2(1.6 \times 10^{-19} C)(0.593 \mu A)(150 \times 10^6 \text{ Hz}) = 2.84 \times 10^{-17} A^2$

$$(b) \sigma_{DB}^2 = 2qI_D B = 2(1.6 \times 10^{-19} \text{ C})(1.0 \text{ nA})(150 \times 10^6 \text{ Hz}) = 4.81 \times 10^{-20} \text{ A}^2$$

$$(c) \sigma_T^2 = \frac{4k_B T}{R_L} B = \frac{4(1.38 \times 10^{-23} \text{ J/K})(293 \text{ K})}{500 \Omega} (150 \times 10^6 \text{ Hz}) = 4.85 \times 10^{-15} \text{ A}^2$$

6-7. Using $R_0 = \frac{\eta q \lambda}{hc} = 0.58 \text{ A/W}$, we have from Eqs. (6-4), (6-11b), (6-15), and (6-17)

$$\left(\frac{S}{N}\right)_Q = \frac{\frac{1}{2}(R_0 P_0 m)^2 M^2}{2qI_p B M^{1/2} M^2} = \frac{R_0 P_0 m^2}{4qB M^{1/2}} = 6.565 \times 10^{12} P_0$$

$$\left(\frac{S}{N}\right)_{DB} = \frac{(R_0 P_0 m)^2}{4qI_D B M^{1/2}} = 3.798 \times 10^{22} P_0^2$$

$$\left(\frac{S}{N}\right)_{DS} = \frac{(R_0 P_0 m)^2 M^2}{4qI_L B} = 3.798 \times 10^{26} P_0^2$$

$$\left(\frac{S}{N}\right)_T = \frac{\frac{1}{2}(R_0 P_0 m)^2 M^2}{4k_B T B / R_L} = 7.333 \times 10^{22} P_0^2$$

where P_0 is given in watts. To convert $P_0 = 10^{-n} \text{ W}$ to dBm, use $10 \log \left(\frac{P_0}{10^{-3}} \right) = 10(3-n) \text{ dBm}$

6-8. Using Eq. (6-18) we have

$$\begin{aligned} \frac{S}{N} &= \frac{\frac{1}{2}(R_0 P_0 m)^2 M^2}{2qB(R_0 P_0 + I_D)M^{5/2} + 2qI_L B + 4k_B T B / R_L} \\ &= \frac{1.215 \times 10^{-16} M^2}{2.176 \times 10^{-23} M^{5/2} + 1.656 \times 10^{-19}} \end{aligned}$$

The value of M for maximum S/N is found from Eq. (6-19), with $x = 0.5$:

$$M_{\text{optimum}} = 62.1.$$

$$6-9. \quad 0 = \frac{d}{dM} \left(\frac{S}{N} \right) = \frac{d}{dM} \left[\frac{\frac{1}{2} I_p^2 M^2}{2q(I_p + I_D)M^{2+x} + 2qI_L + 4k_B T/R_L} \right]$$

$$0 = I_p^2 M - \frac{(2+x)M^{1+x} 2q(I_p + I_D) \frac{1}{2} I_p^2 M^2}{2q(I_p + I_D)M^{2+x} + 2qI_L + 4k_B T/R_L}$$

$$\text{Solving for M: } M_{\text{opt}}^{2+x} = \frac{2qI_L + 4k_B T/R_L}{xq(I_p + I_D)}$$

6-10. (a) Differentiating p_n , we have

$$\frac{\partial p_n}{\partial x} = \frac{1}{L_p} \left(p_{n0} + B e^{-\alpha_s w} \right) e^{(w-x)/L_p} - \alpha_s B e^{-\alpha_s x}$$

$$\frac{\partial^2 p_n}{\partial x^2} = -\frac{1}{L_p^2} \left(p_{n0} + B e^{-\alpha_s w} \right) e^{(w-x)/L_p} + \alpha_s^2 B e^{-\alpha_s x}$$

Substituting p_n and $\frac{\partial^2 p_n}{\partial x^2}$ into the left side of Eq. (6-23):

$$\begin{aligned} & -\frac{D_p}{L_p} \left(p_{n0} + B e^{-\alpha_s w} \right) e^{(w-x)/L_p} + D_p \alpha_s^2 B e^{-\alpha_s x} \\ & + \frac{1}{\tau_p} \left(p_{n0} + B e^{-\alpha_s w} \right) e^{(w-x)/L_p} - \frac{B}{\tau_p} e^{-\alpha_s x} + \Phi_0 \alpha_s e^{-\alpha_s x} \\ & = \left[B \left(D_p \alpha_s^2 - \frac{1}{\tau_p} \right) + \Phi_0 \alpha_s \right] e^{-\alpha_s x} \end{aligned}$$

where the first and third terms cancelled because $L_p^2 = D_p \tau_p$.

Substituting in for B:

$$\text{Left side} = \left[\frac{\Phi_0}{D_p} \frac{\alpha_s L_p^2}{1 - \alpha_s^2 L_p^2} \left(D_p \alpha_s^2 - \frac{1}{\tau_p} \right) + \Phi_0 \alpha_s \right] e^{-\alpha_s x}$$

$$\begin{aligned}
&= \frac{\Phi_0}{D_p} \left[\frac{\alpha_s L_p^2}{1 - \alpha_s^2 L_p^2} \left(\frac{\alpha_s^2 L_p - 1}{\tau_p} \right) + D_p \alpha_s \right] e^{-\alpha_s x} \\
&= \frac{\Phi_0}{D_p} (-D_p \alpha_s + D_p \alpha_s) e^{-\alpha_s x} = 0 \quad \text{Thus Eq. (6-23) is satisfied.}
\end{aligned}$$

$$\begin{aligned}
\text{(b) } J_{\text{diff}} &= q D_p \left(\frac{\partial p_n}{\partial x} \right)_{x=w} \\
&= q D_p \left[\frac{1}{L_p} (p_{n0} + B e^{-\alpha_s w}) - \alpha_s B e^{-\alpha_s w} \right] \\
&= q D_p B \left(\frac{1}{L_p} - \alpha_s \right) e^{-\alpha_s w} + q p_{n0} \frac{D_p}{L_p} \\
&= q \Phi_0 \left(\frac{\alpha_s L_p^2}{1 - \alpha_s^2 L_p^2} \right) \left(\frac{1 - \alpha_s L_p}{L_p} \right) e^{-\alpha_s w} + q p_{n0} \frac{D_p}{L_p} \\
&= q \Phi_0 \frac{\alpha_s L_p}{1 + \alpha_s L_p} e^{-\alpha_s w} + q p_{n0} \frac{D_p}{L_p}
\end{aligned}$$

c) Adding Eqs. (6-21) and (6-25), we have

$$\begin{aligned}
J_{\text{total}} &= J_{\text{drift}} + J_{\text{diffusion}} = q \Phi_0 \left[\left(1 - e^{-\alpha_s w} \right) + \left(\frac{\alpha_s L_p}{1 + \alpha_s L_p} \right) e^{-\alpha_s w} \right] + q p_{n0} \frac{D_p}{L_p} \\
&= q \Phi_0 \left(1 - \frac{e^{-\alpha_s w}}{1 + \alpha_s L_p} \right) e^{-\alpha_s w} + q p_{n0} \frac{D_p}{L_p}
\end{aligned}$$

6-11. (a) To find the amplitude, consider

$$\left(J_{\text{tot}}^* J_{\text{tot}} \right)_{\text{sc}}^{1/2} = q \Phi_0 (S S^*)^{1/2} \quad \text{where } S = \frac{1 - e^{-j\omega t_d}}{j\omega t_d}$$

We want to find the value of ωt_d at which $(S S^*)^{1/2} = \frac{1}{\sqrt{2}}$.

Evaluating $(S S^*)^{1/2}$, we have

$$\begin{aligned}
(S S^*)^{1/2} &= \left[\left(\frac{1 - e^{-j\omega t_d}}{j\omega t_d} \right) \left(\frac{1 - e^{+j\omega t_d}}{-j\omega t_d} \right) \right]^{1/2} \\
&= \frac{\left[1 - \left(e^{+j\omega t_d} + e^{-j\omega t_d} \right) + 1 \right]^{1/2}}{\omega t_d} = \frac{(2 - 2 \cos \omega t_d)^{1/2}}{\omega t_d} \\
&= \frac{[(1 - \cos \omega t_d)/2]^{1/2}}{\omega t_d/2} = \frac{\sin \left(\frac{\omega t_d}{2} \right)}{\frac{\omega t_d}{2}} = \text{sinc} \left(\frac{\omega t_d}{2} \right)
\end{aligned}$$

We want to find values of ωt_d where $(S S^*)^{1/2} = \frac{1}{\sqrt{2}}$.

| x | sinc x | | x | sinc x |
|----------|---------------|--|----------|---------------|
| 0.0 | 1.000 | | 0.5 | 0.637 |
| 0.1 | 0.984 | | 0.6 | 0.505 |
| 0.2 | 0.935 | | 0.7 | 0.368 |
| 0.3 | 0.858 | | 0.8 | 0.234 |
| 0.4 | 0.757 | | 0.9 | 0.109 |

By extrapolation, we find $\text{sinc } x = 0.707$ at $x = 0.442$.

Thus $\frac{\omega t_d}{2} = 0.442$ which implies $\omega t_d = 0.884$

(b) From Eq. (6-27) we have $t_d = \frac{w}{v_d} = \frac{1}{\alpha_s v_d}$. Then

$$\omega t_d = 2\pi f_{3\text{-dB}} t_d = 2\pi f_{3\text{-dB}} \frac{1}{\alpha_s v_d} = 0.884 \quad \text{or}$$

$$f_{3\text{-dB}} = 0.884 \alpha_s v_d / 2\pi$$

6-12. (a) The RC time constant is

$$RC = \frac{R \epsilon_0 K_s A}{w} = \frac{(10^4 \Omega)(8.85 \times 10^{-12} \text{ F/m})(11.7)(5 \times 10^{-8} \text{ m}^2)}{2 \times 10^{-5} \text{ m}} = 2.59 \text{ ns}$$

(b) From Eq. (6-27), the carrier drift time is

$$t_d = \frac{w}{v_d} = \frac{20 \times 10^{-6} \text{ m}}{4.4 \times 10^4 \text{ m/s}} = 0.45 \text{ ns}$$

$$\text{c) } \frac{1}{\alpha_s} = 10^{-3} \text{ cm} = 10 \text{ } \mu\text{m} = \frac{1}{2} w$$

Thus since most carriers are absorbed in the depletion region, the carrier diffusion time is not important here. The detector response time is dominated by the RC time constant.

6-13. (a) With $k_1 \approx k_2$ and k_{eff} defined in Eq. (6-10), we have

$$(1) \quad 1 - \frac{k_1(1 - k_1)}{1 - k_2} = 1 - \frac{k_1 - k_1^2}{1 - k_2} \approx 1 - \frac{k_2 - k_1^2}{1 - k_2} = 1 - k_{\text{eff}}$$

$$(2) \quad \frac{(1 - k_1)^2}{1 - k_2} = \frac{1 - 2k_1 + k_1^2}{1 - k_2} \approx \frac{1 - 2k_2 + k_1^2}{1 - k_2}$$

$$= \frac{1 - k_2}{1 - k_2} - \frac{k_2 - k_1^2}{1 - k_2} = 1 - k_{\text{eff}}$$

Therefore Eq. (6-34) becomes Eq. (6-38):

$$F_e = k_{\text{eff}} M_e + 2(1 - k_{\text{eff}}) - \frac{1}{M_e}(1 - k_{\text{eff}}) = k_{\text{eff}} M_e + \left(2 - \frac{1}{M_e}\right)(1 - k_{\text{eff}})$$

(b) With $k_1 \approx k_2$ and k'_{eff} defined in Eq. (6-40), we have

$$(1) \quad \frac{k_2(1 - k_1)}{k_1^2(1 - k_2)} \approx \frac{k_2 - k_1^2}{k_1^2(1 - k_2)} = k'_{\text{eff}}$$

$$(2) \quad \frac{(1 - k_1)^2 k_2}{k_1^2(1 - k_2)} = \frac{k_2 - 2k_1 k_2 + k_2 k_1^2}{k_1^2(1 - k_2)}$$

$$\approx \frac{(k_2 - k_1^2) - (k_1^2 - k_2 k_1^2)}{k_1^2(1 - k_2)} = k'_{\text{eff}} - 1$$

Therefore Eq. (6-35) becomes Eq. (6-39): $F_h = k'_{\text{eff}} M_h - \left(2 - \frac{1}{M_h}\right)(k'_{\text{eff}} - 1)$

- 6-14. (a) If only electrons cause ionization, then $\beta = 0$, so that from Eqs. (6-36) and (6-37), $k_1 = k_2 = 0$ and $k_{\text{eff}} = 0$. Then from Eq. (6-38)

$$F_e = 2 - \frac{1}{M_e} \approx 2 \text{ for large } M_e$$

- (b) If $\alpha = \beta$, then from Eqs. (6-36) and (6-37), $k_1 = k_2 = 1$ so that

$k_{\text{eff}} = 1$. Then, from Eq. (6-38), we have $F_e = M_e$.