

Signals and Spectra (3)

6-1, 6-2

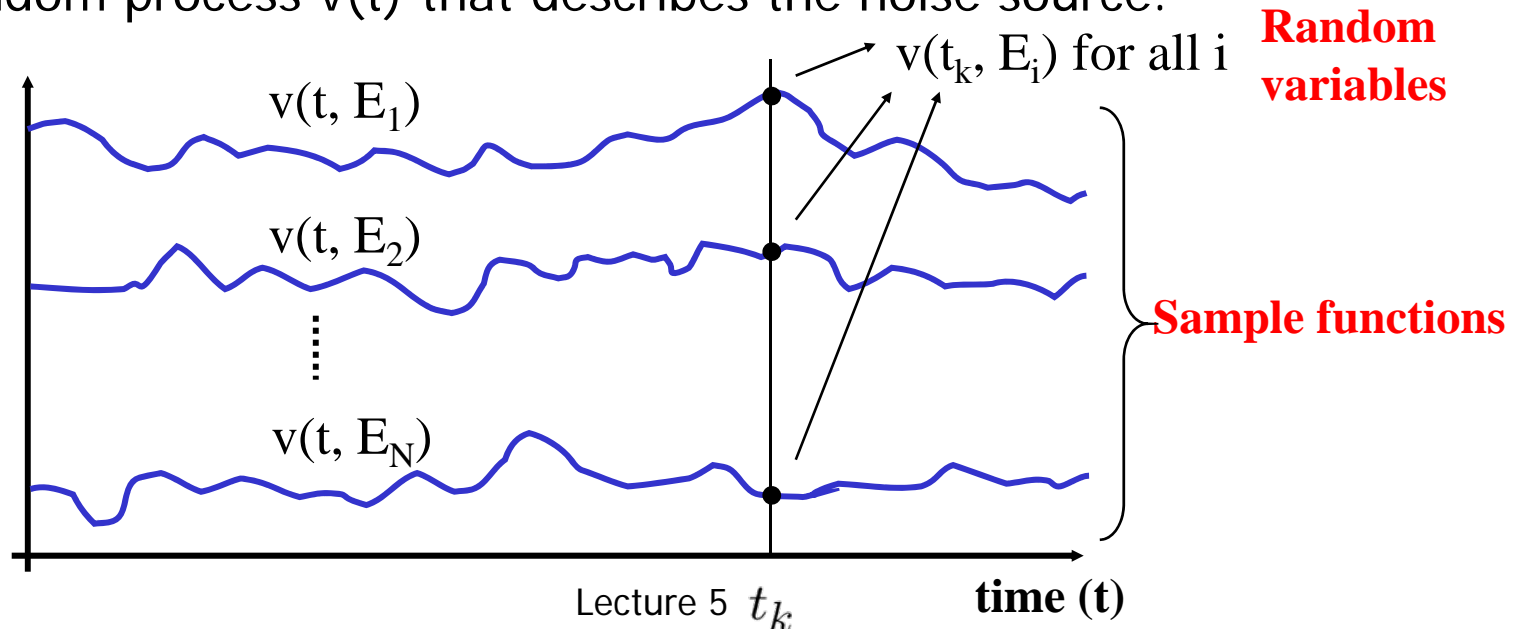
Lecture 5, 2008-09-19

Introduction

- Random Process
- Stationary
- Ergodic
- Correlation Functions
- Power Spectral Density
- Wiener-Khintchine Theorem

Random Process

- A real **random process** (or stochastic process) is an indexed set of real functions of some parameter (usually time) that has certain statistical properties.
- Imagine voltage waveforms that might be emitted from a noise source. In general, $v(t, E_i)$ denotes the waveform that is obtained when the event E_i of the sample space occurs. $v(t, E_i)$ is said to be a sample function of the sample space. The set of all possible sample functions $\{v(t, E_i)\}$ is called the ensemble and defines the random process $v(t)$ that describes the noise source.



Random Process

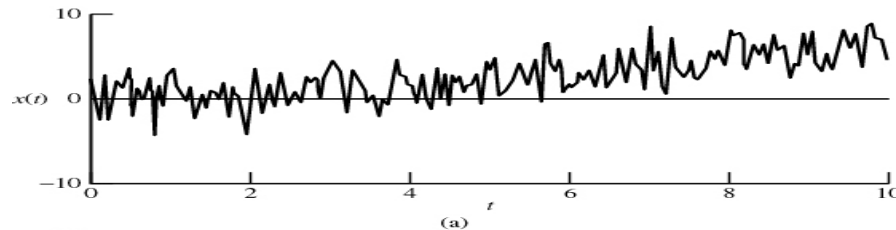
- The difference between a random variable and a random process is that for a random variable, an outcome in the sample space is mapped into a number, whereas for a random process it is mapped into a function of time.
- A random process may be described by an indexed set of random variables.
- To describe a general random process $x(t)$ completely, an N-dimensional PDF, $f_x(\mathbf{x})$, is required, where $\mathbf{X} = (x_1, x_2, \dots, x_N)$, $x_j = x(t_j)$ and $N \rightarrow \infty$. $f_x(\mathbf{x}) = f_x(x(t_1), x(t_2), \dots, x(t_N))$

Stationary

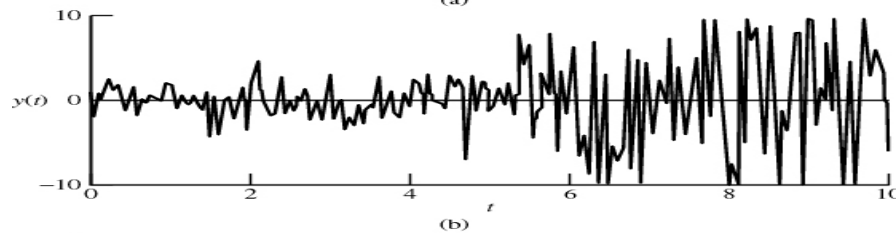
- A random process $x(t)$ is said to be **stationary** to the order N if, for any t_1, t_2, \dots, t_N

$$f_x(x(t_1), x(t_2), \dots, x(t_N)) = f_x(x(t_1 + t_0), x(t_2 + t_0), \dots, x(t_N + t_0))$$

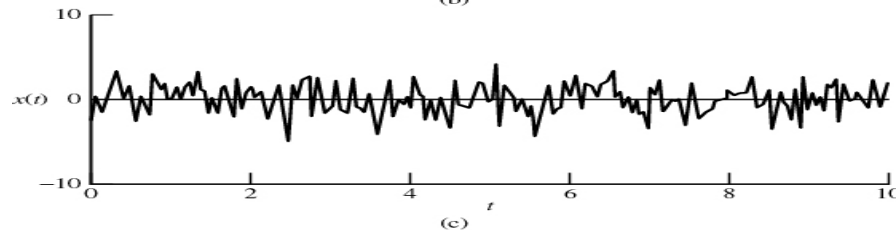
Where t_0 is any arbitrary real constant. Furthermore, the process is said to be strictly stationary if it is stationary to the order $N \rightarrow \infty$.



(a) Time-varying mean



(b) Time-varying variance



(c) Stationary

Example

■ First-order stationary

The requirement for first-order stationary is that is the first-order PDF not be a function of time. Let the random process be $x(t) = A \sin(\omega t + \theta)$

Case 1: Stationary Result. Assume A and ω are deterministic constants and θ is a random variable uniformly distributed over $-\pi$ to π .

$$f(x) = \begin{cases} \frac{1}{\pi \sqrt{A^2 - x^2}}, & |x| \leq A \\ 0, & |x| > A \end{cases}$$

Case 2: Nonstationary Result. Assume A , ω and θ are deterministic constants.

$$f(x) = \delta(x - A \sin(\omega t + \theta))$$

Ergodic

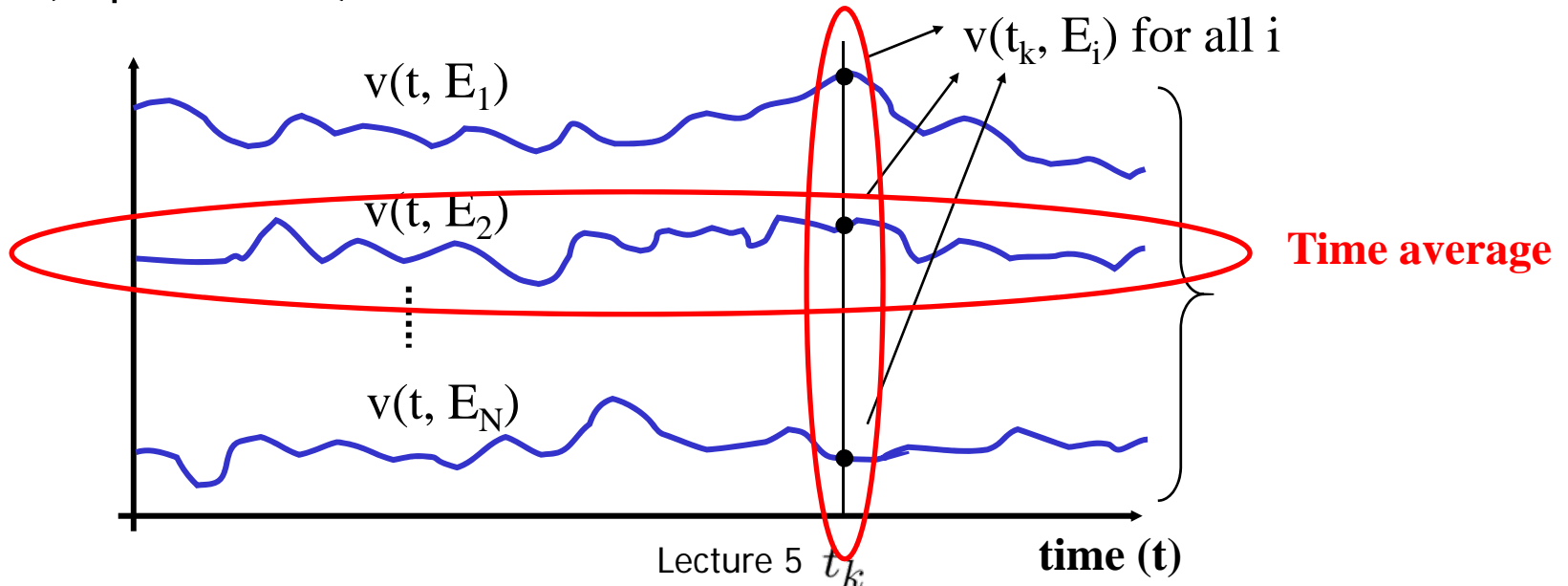
- The time average is

$$\langle [x(t)] \rangle = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} [x(t)] dt$$

- The ensemble average is

$$\overline{[x(t)]} = \int_{-\infty}^{\infty} [x] f_x(x) dx = m_x$$

- A random process is said to be **ergodic** if all time averages of any sample function are the corresponding ensemble averages (expectations).



Example

- Let a random process be given by $x(t) = A \cos(\omega t + \theta)$, Assume A and ω are deterministic constants and θ is a random variable uniformly distributed over 0 to 2π .

- The ensemble averages

$$\bar{x} = \int_{-\infty}^{\infty} [x(\theta)] f_{\theta}(\theta) d\theta = \int_0^{2\pi} [A \cos(\omega t + \theta)] \frac{1}{2\pi} d\theta = 0$$

- The time averages

$$\langle x(t) \rangle = \frac{1}{T_0} \int_0^{T_0} A \cos(\omega t + \theta) dt = 0$$

- The time average is equal to the ensemble average.
- Shall we conclude that the process is ergodic?
- No. Because we have not evaluated all the possible time and ensemble averages. In general, it is difficult to prove that a process is ergodic

Ergodic vs. Stationary

- The ergodic process must be stationary, because otherwise the ensemble averages would be a function of time.
- However, if a process is known to be stationary, it may or may not be ergodic.
- Excise 6.2

Correlation Functions

- The **autocorrelation** function of a random process $x(t)$ is

$$R_x(t_1, t_2) = \overline{x(t_1)x(t_2)} = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x_1 x_2 f_x(x_1, x_2) dx_1 dx_2$$

where $x_1 = x(t_1)$ and $x_2 = x(t_2)$.

- If the process is stationary to the second order, the autocorrelation function is a function only of the time difference $\tau = t_2 - t_1$

$$R_x(\tau) = \overline{x(t)x(t+\tau)}$$

if $x(t)$ is second-order stationary

- A random process is said to be **wide-sense stationary** if

$$\frac{d\overline{x(t)}}{dt} = 0 \text{ and } R_x(t_1, t_2) = R_x(t_2 - t_1)$$

- The auto-correlation function has following properties:

$$R_x(0) = \overline{x^2(t)}$$

$$R_x(\tau) = R_x(-\tau)$$

$$R_x(0) \geq |R_x(\tau)|$$

Cross-Correlation Function

- The **cross-correlation** function of two random process $x(t)$ and $y(t)$ is

$$R_{xy}(t_1, t_2) = \overline{x(t_1)y(t_2)}$$

- If $x(t)$ and $y(t)$ are jointly stationary, the cross-correlation function is

$$R_{xy}(t_1, t_2) = R_{xy}(t_2 - t_1)$$

- The cross-correlation function has following properties:

$$R_{xy}(-\tau) = R_{yx}(\tau)$$

$$|R_{xy}(\tau)| \leq \sqrt{R_x(0)R_y(0)}$$

$$|R_{xy}(\tau)| \leq \frac{1}{2} [R_x(0) + R_y(0)]$$

- Two random processes $x(t)$ and $y(t)$ are said to be uncorrelated if

$$R_{xy}(\tau) = \overline{[x(t)][y(t+\tau)]} = m_x m_y \quad \text{for all values of } \tau$$

- Two random processes $x(t)$ and $y(t)$ are said to be orthogonal if

$$R_{xy}(\tau) = 0 \quad \text{for all values of } \tau$$

Correlation Measurement



Power Spectral Density

- The **power spectral density** for a random process $x(t)$ is given by

$$P_x(f) = \lim_{T \rightarrow \infty} \left(\frac{\overline{|X_T(f)|^2}}{T} \right)$$

where

$$X_T(f) = \int_{-T/2}^{T/2} x(t) e^{-j2\pi ft} dt$$

Wiener-Khintchine Theorem

- When $x(t)$ is a wide-sense stationary process, the PSD can be obtained from the Fourier transform of the autocorrelation function

$$P_x(f) = \int_{-\infty}^{\infty} R_x(t) e^{-j2\pi ft} dt$$

Conversely,

$$R_x(t) = \int_{-\infty}^{\infty} P_x(f) e^{j2\pi ft} df$$

- To calculate the PSD of a random process:
 1. Direct method: Use the original definition of PSD
 2. Indirect method: Use Wiener-Khintchine Theorem

Proof of W-K Theorem

From the definition of PSD

$$P_x(f) = \lim_{T \rightarrow \infty} \left(\frac{\overline{|X_T(f)|^2}}{T} \right)$$

Where

$$\begin{aligned} \overline{|X_T(f)|^2} &= \overline{\left| \int_{-T/2}^{T/2} x(t) e^{-j2\pi ft} dt \right|^2} \\ &= \int_{-T/2}^{T/2} \int_{-T/2}^{T/2} \overline{x(t_1) x(t_2)} e^{-j2\pi f(t_1 - t_2)} dt_1 dt_2 \end{aligned}$$

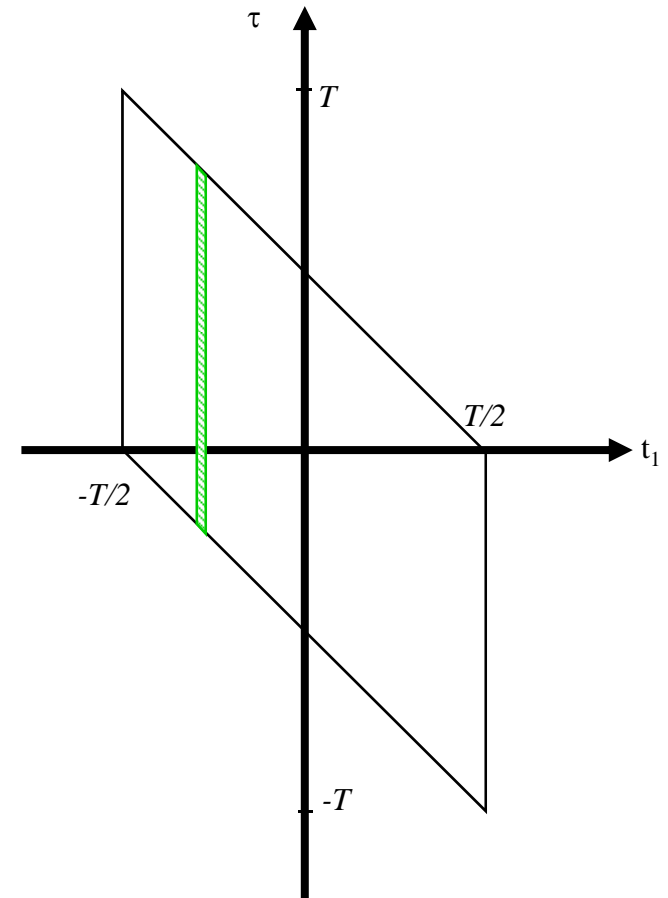
Proof of W-K Theorem

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Region of Integration

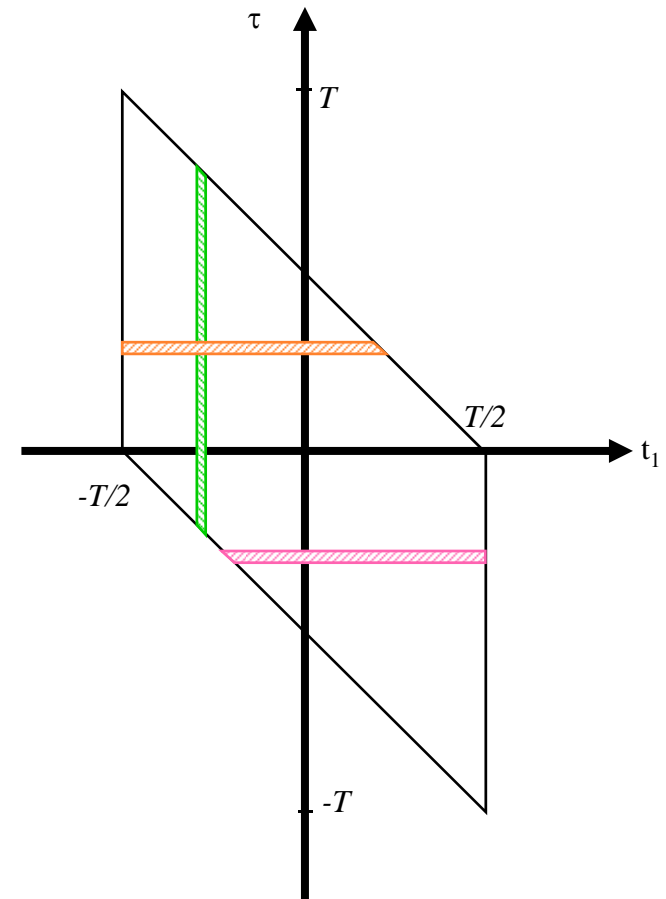
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From the definition of PSD

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Where

$$\begin{aligned} \overline{|X_T(f)|^2} &= \overline{\left| \int_{-T/2}^{T/2} x(t) e^{-j2\pi f t} dt \right|^2} \\ &= \int_{-T/2}^{T/2} \int_{-T/2}^{T/2} \overline{x(t_1) x(t_2) e^{-j2\pi f (t_1 - t_2)}} dt_1 dt_2 \\ &= \int_{-T/2}^{T/2} \left[\int_{-T/2 - t_1}^{T/2 - t_1} R_x(t_1, t_1 + \tau) e^{-j2\pi f \tau} d\tau \right] dt_1 \\ &= \int_{-T}^0 \left[\int_{-T/2 - \tau}^{T/2} R_x(t_1, t_1 + \tau) e^{-j2\pi f \tau} dt_1 \right] d\tau \\ &+ \int_0^T \left[\int_{-T/2}^{T/2 - \tau} R_x(t_1, t_1 + \tau) e^{-j2\pi f \tau} dt_1 \right] d\tau \end{aligned}$$



Region of Integration

Cont'

If $x(t)$ is stationary,

$$\begin{aligned} \overline{|X_T(f)|^2} &= \int_{-T}^0 R_x(\tau) e^{-j2\pi f\tau} \left[t_1 \Big|_{-T/2-\tau}^{T/2} \right] d\tau + \int_0^T R_x(\tau) e^{-j2\pi f\tau} \left[t_1 \Big|_{-T/2}^{T/2-\tau} \right] d\tau \\ &= \int_{-T}^T (T - |\tau|) R_x(\tau) e^{-j2\pi f\tau} d\tau \end{aligned}$$

Then

$$\begin{aligned} P_x(f) &= \lim_{T \rightarrow \infty} \int_{-T}^T \left(1 - \frac{|\tau|}{T} \right) R_x(\tau) e^{-j2\pi f\tau} d\tau \\ &= \int_{-\infty}^{\infty} R_x(\tau) e^{-j2\pi f\tau} d\tau \end{aligned}$$

Example

■ example 6-3

Properties of PSD

- $P_x(f)$ is always real
- $P_x(f) \geq 0$
- When $x(t)$ is real, $P_x(-f) = P_x(f)$
- Total normalized power $\int_{-\infty}^{\infty} P_x(f)df = P$
- $P_x(0) = \int_{-\infty}^{\infty} R_x(\tau)d\tau$

General Formula for the PSD of Digital Signals

$$P_x(f) = \frac{|F(f)|^2}{T_s} \left[\sum_{-\infty}^{\infty} R(k) e^{jk\omega T_s} \right]$$

$$R(k) = \overline{a_n a_{n+k}} = \sum_{i=1}^I (a_n a_{n+k})_i P_i$$

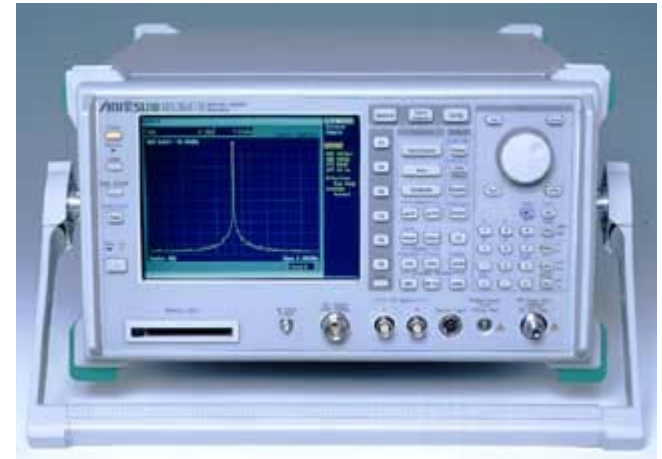
- P_i is the probability of getting the product $(a_n a_{n+k})_i$, of which there are I possible values
- $F(f)$ is the spectrum of the pulse shape of the digital symbol

White-Noise Processes

- A random process $x(t)$ is said to be a white-noise process if the PSD is constant over all frequencies, $P_x(f) = N_0/2$, where N_0 is a positive constant.
- The autocorrelation function for the white noise is obtained by taking the inverse Fourier transform $R_x(t) = N_0/2 \delta(t)$
- Any two different samples of a white noise process are uncorrelated. Since thermal noise is a Gaussian process and the samples are uncorrelated, the noise samples are also independent.
- The effect on the detection process of a channel with Additive White Gaussian Noise is that the noise affects each transmitted symbol independently. Memoryless channel.

Measurement of PSD

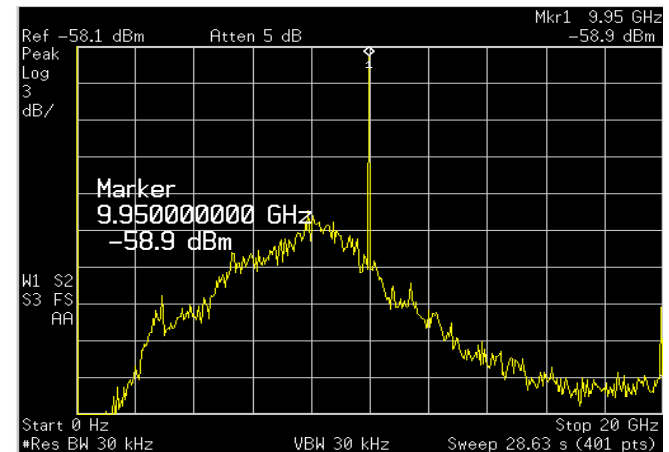
- Analog techniques
 - Narrowband filters
 - RF spectrum analyzer
- Numerical computation



$$P_T(f) = \frac{|X_T(f)|^2}{T}$$

$$\overline{P_T(f)} = \left[\frac{|X_T(f)|^2}{T} \right] = F \left[R_x(\tau) \Lambda \left(\frac{\tau}{T} \right) \right]$$

$$= TP_x(f) * \left(\frac{\sin \pi fT}{\pi fT} \right)^2$$



Homework

- 6-2, 6-3, 6-9, 6-15, 6-17

